CHAPTER 4

THE EFFECT OF EXTERNAL NOISE

4.1 Introduction

When an external noise $n_i(t)$ is added to the input signal x(t), then the input of the linear system is

$$x_{i}(t) = x(t) + n_{i}(t)$$
 (4.1)

The output response y_i(t) of this system can be written as

$$y_{i}(t) = \int_{0}^{\infty} h(u)x_{i}(t-u)du \qquad (4.2)$$

The crosscorrelation between the input x(t) and the output response $y_i(t)$ becomes

$$\boldsymbol{y}_{\mathbf{x}\mathbf{y}_{\mathbf{i}}}(\tau) = \frac{1}{T} \int_{0}^{T} \mathbf{x}(t) \mathbf{y}_{\mathbf{i}}(t+\boldsymbol{\tau}) dt \qquad (4.3)$$

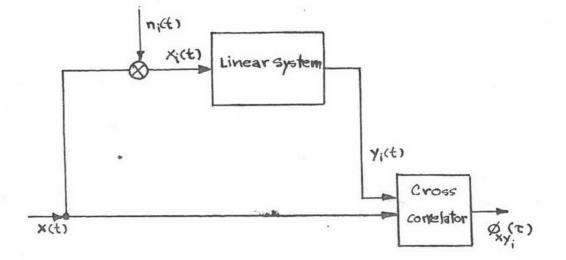
From eqns. (4.2) and (4.3), we obtain

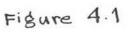
$$\mathscr{D}_{xy_{i}}(\tau) = \frac{1}{T} \int_{\mathbf{0}} \int_{\mathbf{0}}^{\tau} h(u)x(t)x_{i}(t+\tau-u)dudt \qquad (4.4)$$

From eqns. (4.1) and (4.4), we have

$$\varphi_{xy_{i}}(\tau) = \int_{0}^{\infty} h(u) \varphi_{xx}(\tau-u) du + \int_{0}^{\infty} h(u) \varphi_{xn_{i}}(\tau-u) du (4.5)$$

It can be seen that the first integral term on the righthand side of eqn. (4.5) is the same as eqn. (1.3), therefore the second integral term must be evaluated, and for this case the crosscorrelation function $\mathscr{G}_{\mathrm{Xn}_{i}}(\mathcal{T})$ is required. However there are





The cross correlation between input signal x(t) and output response y; (t)

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two interested forms of the external noise to be considered, they are the time polynomial function and the white noise signals.

4.2 Determination of $\phi_{xn}(\tau)$ Effected by the Time Polynomial

Function Noise Signal

The time polynomial function signal $n_i(t)$ may be written in the form

$$n_{i}(t) = b_{0} + b_{1} t + b_{2} t^{2} + \dots + b_{m} t^{m}$$

$$= \sum_{i=0}^{m} b_{i} t^{i}$$
(4.6)
(4.7)

where b₀, b₁, b₂, ..., b_m are coefficients of time varible t of the order 0,1,2,...,m respectively, and m is the positive integer.

Since the input signal x(t) is periodic with period T, the crosscorrelation function $\emptyset_{xn}(\tau)$ will be determined over the same period. Hence, we have

$$\varphi_{\mathbf{xn}_{\mathbf{i}}}(\tau) = \frac{1}{T} \int_{\mathbf{0}}^{\tau} \mathbf{x}(t) \mathbf{n}_{\mathbf{i}}(t+\tau) dt \qquad (4.8)$$

From eqns. (4.7) and (4.8), we obtain

$$\emptyset_{xn_{i}}(\tau) = \frac{1}{T} \int_{0}^{1} x(t) \sum_{i=0}^{m} b_{i}(t+\tau)^{i} dt \qquad (4.9)$$

Rearrange the eqn. (4.9), (see Appendix E), we have

$$\emptyset_{xn_{i}}(\mathcal{T}) = \frac{1}{T} \int_{0}^{T} x(t) \sum_{i=0}^{m} t^{i} \sum_{j=i}^{m} \mathbf{j} \mathbf{c}_{i} \mathbf{b}_{j} \mathbf{1}^{j-i} dt$$

$$\text{where } \mathbf{j}_{i} = \frac{j!}{(j-i)!i!}$$

$$(4.10)$$

$$\begin{split} \boldsymbol{\emptyset}_{\mathbf{xn_{i}}}(\tau) &= \sum_{\substack{i=0 \ o \\ \mathbf{xn_{i}}}}^{m} \frac{1}{T} \int_{\mathbf{x}(t)}^{t} t^{i} dt \sum_{\substack{j=i \\ \mathbf{j=i}}}^{m} j_{\mathbf{C}_{i}b_{j}} \tau^{j-i} \\ &= \sum_{\substack{i=0 \ i \\ \mathbf{j=i}}}^{m} M_{i} \sum_{\substack{j=i \\ \mathbf{j=i}}}^{m} j_{\mathbf{C}_{i}b_{j}} \tau^{j-i} \end{split}$$
(4.11)

where M_i is the time moment of the input signal $x(t) = \frac{1}{T_o} \int_{0}^{\infty} x(t) t^i dt$ Since the eqn. (4.11) is not convenient to determine the crosscorrelation function $\mathscr{O}_{xn_i}(\mathcal{T})$. Hence, it may be rearranged as (see also Appendix E)

$$\emptyset_{\mathbf{xn_{i}}}(\tau) = A_{0} + A_{1}\tau + A_{2}\tau^{2} + \dots + A_{m}\tau^{m}$$
(4.12)

$$A_{\mathbf{i}} = \sum_{\substack{j=\mathbf{i}\\ j=\mathbf{i}}}^{m} {}^{\mathbf{j}C_{\mathbf{i}}} \mathbf{b}_{\mathbf{j}}^{\mathbf{M}} \mathbf{j} - \mathbf{i}$$

$$M_{\mathbf{j}-\mathbf{i}} = \frac{1}{T} \int_{0}^{\tau} \mathbf{x}(\mathbf{t}) \mathbf{t}^{\mathbf{j}-\mathbf{i}} d\mathbf{t}$$

In practice, it is simpler to delay the input signal $\mathbf{x}(\mathbf{t})$

where

In practice, it is simpler to delay the input signal x(t)rather than the external noise $n_i(t)$. The crosscorrelation between the input signal and the external noise can be expressed as

$$\mathscr{D}_{\mathrm{xn}_{\mathbf{i}}}(\tau) = \frac{1}{T} \int_{0}^{T} \mathbf{x}(t-\tau) \mathbf{n}_{\mathbf{i}}(t) dt \qquad (4.13)$$

By replacing $t = t + \mathcal{T}$, it can be shown that eqn. (4.13) is reduced to eqn. (4.8).

From eqns. (4.7) and (4.13), we obtain

$$\begin{split} \mathscr{P}_{\mathbf{x}\mathbf{n}_{\mathbf{i}}}(\tau) &= \frac{1}{T} \int_{\mathbf{x}}^{T} (\mathbf{t} - \tau) \sum_{i=0}^{m} \mathbf{b}_{\mathbf{i}} \mathbf{t}^{\mathbf{i}} d\mathbf{t} \\ \mathscr{P}_{\mathbf{x}\mathbf{n}_{\mathbf{i}}}(\tau) &= \sum_{i=0}^{m} \mathbf{b}_{\mathbf{i}} \left\{ \frac{1}{T} \int_{\mathbf{v}}^{T} \mathbf{x}(\mathbf{t} - \tau) \mathbf{t}^{\mathbf{i}} d\mathbf{t} \right\} \\ &= \mathbf{b}_{0}^{\mathbf{M}_{0}}(\tau) + \mathbf{b}_{1}^{\mathbf{M}_{1}}(\tau) + \dots + \mathbf{b}_{m}^{\mathbf{M}_{m}}(\tau) \quad (4.14) \end{split}$$

where $M_i(\tau) = \frac{1}{T} \int_{0}^{1} x(t-\tau)t^i dt$, is the new time moment of x(t). It is seen that the result of $\emptyset_{xn_i}(\tau)$ in eqn. (4.14) which

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is the modified form of $\mathscr{P}_{xn_i}(\mathcal{T})$ in eqn.(4.8), will give the same result as obtained by eqn.(4.12).

4.3 Determination of $\emptyset_{xn}(\mathbf{\hat{c}})$ Effected by the White Noise Signal

Normally, it is difficult to predict the waveform of the white noise, but its properties are well-known. Thus the crosscor-relation function $\emptyset_{xn}(\tau)$ may be determined as follow: From eqn.(4.8), it can be rewritten as

$$\varphi_{\mathrm{xn}_{i}}(\tau) = \frac{1}{T} \int_{0}^{1} x(t) n_{i}(t+\tau) dt \qquad (4.15)$$

The square value of $\mathscr{O}_{\mathrm{xn}_{i}}(\mathcal{T})$ is

$$\varphi_{xn_{i}}^{2}(\tau) = \left\{ \frac{1}{T} \int_{0}^{\tau} x(t)n_{i}(t+\tau) dt \right\}^{2}$$
(4.16)

and see Appendix F, we have

$$\phi_{\mathrm{xn}_{i}}^{2}(\mathcal{L}) = \phi_{\mathrm{xx}}(\mathcal{L}) \phi_{\mathrm{n}_{i}\mathrm{n}_{i}}(\mathcal{L})$$
(4.17)

Since the autocorrelation function of the white noise $n_i(t)$ is known to be KS(0), the eqn.(4.17) becomes

Substituting for the value of $\beta_{xx}(2)$ from property h, we have

Thus the value of $\emptyset_{xn_i}(\gamma)$ is

$$\emptyset_{\mathrm{xn}_{\mathbf{i}}}(\mathbf{\hat{\tau}}) = a\sqrt{K} , \text{ for } \mathbf{\hat{\tau}} = 0$$

$$= 0 , \text{ for } \mathbf{\hat{\tau}} \neq 0$$

$$\left. \right\}$$

$$(4.20)$$

or it can be rewritten in the form

The approximation value of $\emptyset_{xn_i}(\mathcal{T})$ will be assumed constant for all range of \mathcal{T} as its peak value at $\mathcal{T} = 0$, thus

$$\mathscr{P}_{xn_i}(\mathcal{T}) \stackrel{:}{:} a\sqrt{K}$$
, for all range of \mathcal{T} (4.22)

In this case, the error due to the above assumption, for $2 \neq 0$, is neglected when the signal-to-noise ratio is large. The modified form of $\beta_{xn}(2)$ for the white noise will have the same result as shown in eqn.(4.22)