#### CILIPTER II.

DESCRIPTION OF THE ANALOGUE COMPUTER

The variable quantities of a physical problem arcrelated in various ways. An analogue computer operates by representing these variables by D.C. voltages, and the eircuit is so constructed that the relations between the 2.3. voltages are the same as the relations between the physical variables. These relationships are usually in the form of differential equations. In order to understand how a difforential equation is solved on an analogue computer, it is necessary study in some detail the basic computing components. These computing components are the building blocks in forming any model and their characteristics are explained in sufficient detail below to enable the reader to follow the discussion in Chapter IV.

# (2.1) Operational Amplifiers (1), (2)

An operational amplifier, sometimes called a D.C. amplifier, which performs the basic mathematical operations necessary for the solution of the probleme is the heart of the modern analogue computer. Operational amplifiers used in an analogue computer are usually constructed with vacuumtubes of transistors and are voltage amplifiers, having an output voltage range \$400 volts for vacuum-tube and \$40 volta

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for transistorized. The D.C. amplifier used in an analogue computer is a high gain direct-coupled amplifier with negative feedback. In practice, the forward gain of the amplifier, denoted by -A, has a value in the range of  $10^{2}-10^{8}$  for all expected computer operations. The gain of an amplifier is given by

where  $e_g$  represents the grid voltage, and  $e_0$  the output voltage of the amplifier. (See Fig.2.1).

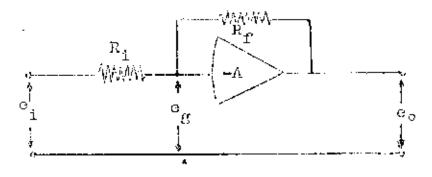


Fig.2.1: The D.C.-eperational amplifier

In Pig.2.1, of represents the input voltage,  $R_{\rm f}$  the feedback resistor and  $R_{\rm i}$  the input resistor.

From equation (2.1), if -A approaches  $-\infty$ , ogen approaches zero. In practice this conditon is realized by using high-gain amplifiers. The linear mathematical operations are performed by using a high-gain D.C. amplifier, as shown in equation (2.2)

$$c_0 = -\frac{R_2}{R_1} c_1$$
 (2.2)

which can be written  $c_0 = -K c_1$ , where  $K = \frac{R_1}{R_1}$ . This amounts to multiplication by a constant coefficient and this ratio  $\frac{R_1}{R_2}$  is comparatively small (less than 100).

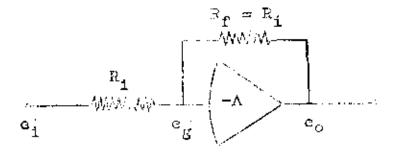
From equation (2.2), if  $R_{f} > R_{i}$  then  $e_{0} > -e_{i}$  and there is a voltage gain. If  $R_{f} = R_{i}$ , then  $e_{0} = -e_{i}$  and the amplifier becomes a sign-changer (also called an invertor). Hence any desired gain in equation (2.2) can be produced simply by choosing the correct elements for  $R_{i}$  and  $R_{f}$ . Furthermore amplifiers are copable of performing, in addition, the operations of: (a)

- a) Hultiplication by -1. or sign-changer
- b) Multiplication by a constant
- c) Addition br summation
- d) Integration ( $n_{\varphi}$  is replaced by a capacitor)

a) Operational Amplificr as Sign-Changer or Invertor.

From equation (2.2), if  $R_f = R_i$  the output voltage will be equal but opposite in sign to the input voltage. We obtained

Thus the operational amplifier can be used as a signchanger or an inverter. The circuit of the sign-changer is represented in Fig.2.2.



lig.2.2 Operational amplifier as a sign-changer.

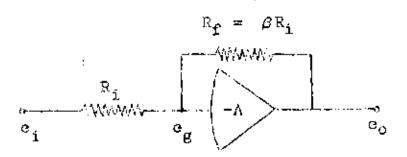
## b) Operational Amplifier for multiplying by

#### a constant

If the value of the feedback resistor is  $\beta$  times the value of the input resistor, then the output voltage will be  $\beta$  times the input voltage. From equation (2.2) ...when  $R_{\rm f} = \beta R_{\rm f}$ , where  $\beta > 1$ , we have

 $c_0 = -\beta c_1$ , where  $\beta > 1$  ----- (2.4)

The sign-changer circuit is shown in Fig.2.3



<u>Fig.2.3</u> Operational Amplifier for Multiplication by  $\beta$ 

Division by a constant  $\beta$  can be treated as multiplication using  $\frac{4}{\beta}$ . The relation between the input voltage and the output voltage is represented in equation (2.5).

$$o_0 = -\frac{1}{9}o_i$$
, where  $\beta > 1$  ----- (2.5)

#### c) Cperational Amplifier as Adder or Summer.

Nore than one input can be applied to the operational amplifier. In this case, the operational amplifier becomes an adder or summer.

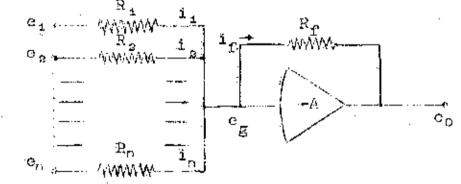


Fig.2.19 Operational Amplifier as Adder.

The inclusion of more than one input resistor to a high gain d-c amplifier circuit, each resistor having a voltage applied to it, changes the input-output relationship to:

$$\frac{\mathbf{C}_{\mathbf{n}}}{\mathbf{R}_{\mathbf{1}}} = -\left[\frac{\mathbf{C}_{\mathbf{1}}}{\mathbf{R}_{\mathbf{2}}} + \frac{\mathbf{C}_{\mathbf{2}}}{\mathbf{R}_{\mathbf{2}}} + \frac{\mathbf{C}_{\mathbf{n}}}{\mathbf{R}_{\mathbf{3}}} + \dots + \frac{\mathbf{C}_{\mathbf{n}}}{\mathbf{R}_{\mathbf{n}}}\right]$$

[That is to say the current  $i_{\Gamma}$  flowing through with feedback resistor must be the algebraic sum of currents  $i_{4}$ ,

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 $i_2, \ldots, i_n$ , flowing through the input resistors since the amplifier input voltage and current are zero. Thus:

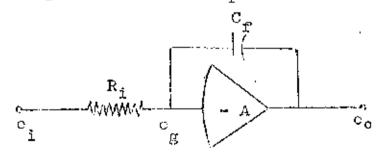
$$e_{0} = -\left[\frac{R_{f}}{R_{t}}e_{1} + \frac{R_{f}}{R_{s}}e_{2} + \dots + \frac{R_{f}}{R_{n}}e_{n}\right]$$
  

$$\therefore e_{0} = -\frac{R}{R_{t}}K_{k}e_{k} - \dots + (2.6)$$
  
where  $K_{k} = \frac{R_{f}}{R_{k}}$ ,  $k = 1$ ,  $2$ ,  $\dots$ , and the  $e_{k}$  are the input voltages.]

Hence the operational amplifier can be used to add. In equation (2.6), by using equal values for all resistors, one obtains a simple algebraic summation with the usual inversion associated with every computing amplifier. If resistors have different values, then each input voltage is multiplied by a factor, given by the ratio of the feedback resistor to the input resistor, before it is added to the sum.

### d) Operational Amplifier as Integrator.

Integration is performed by replacing the feedback resistor  $R_{f}$  by a capacitor  $C_{p}$  as shown in Fig.2.5.



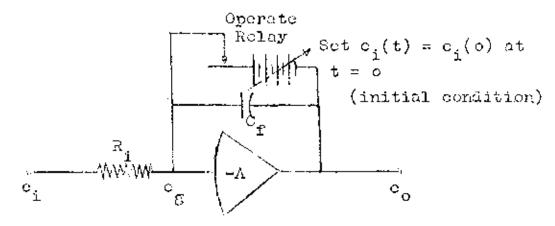
<u>Fig.2.5</u> Integrator with single-input.

The mathematical relation between the input voltage and the output voltage of the operational amplifier used as an integrator is shown in equation (2.7):

$$v_0(t) = -\frac{1}{R_1 C_f} \int_0^t v_1(t) dt + v_1(0) ---- (2.7)$$

where  $e_i(o)$  is the constant of integration (initial condition) and is the voltage across the foodback capacitor  $C_{\underline{r}}$  at t = 0. Thus the operational amplifier can integrate.

It is necessary to be able to control the operation of integration and also to be able to apply an initial charge to the especitor to set the initial value  $e_i(o)$  of the output voltage  $e_0$ . This is done by connecting a relay<sup>(a),(a)</sup> minimizial condition power supply across the amplifier as shown in Fig.2.6. Initially the relay switch is closed. It is then opened to carry out the computation.



<u>Fig.2.6</u>: Integrator with Operate Relay and Initial Condition.

Several voltages can be connected to the input of the integrating amplifier, as shown in Nig.2.7.

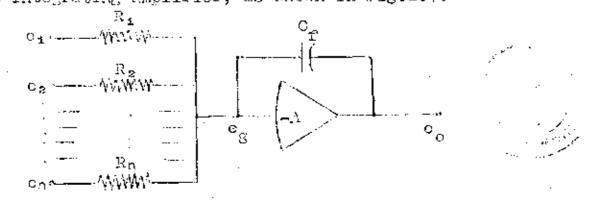


Fig.2.7: Kultiple inputs integrator

The output voltage of the integrating amplifier for multiple inputs is:

$$c_{0}(t) = -\frac{1}{C_{f}} \int_{0}^{t} \left( \frac{c_{1}(t)}{R_{1}} + \frac{c_{2}(t)}{R_{2}} + \cdots \right) + \frac{c_{n}(t)}{R_{n}} dt + c_{0}(0).$$

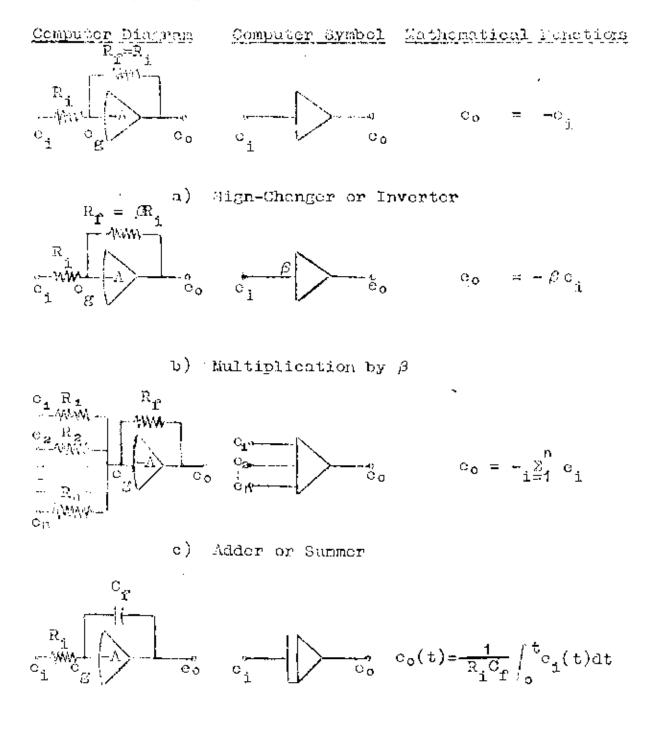
If we let  $\frac{1}{C_{\Gamma}R_{i}} = K_{i}$ , then we obtain  $e_{o}(t) = -\int_{0}^{t} \sum_{i=1}^{n} K_{i}e_{i}(t)dt + e_{0}(0) - (2.8)$ 

where  $o_0(o)$  is the constant of integration.

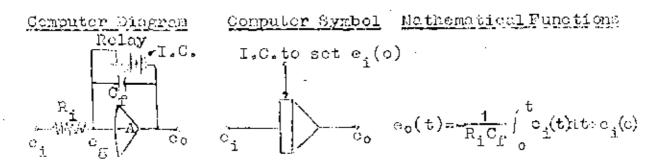
Furthermore, an amplifier may be used to differentiate with respect to t, but since this operation magnifies errors and fluctuates in the computer voltage, it is rarely used.

The symbols frequently used to represent the various functions of an operational amplifier necessary to selve

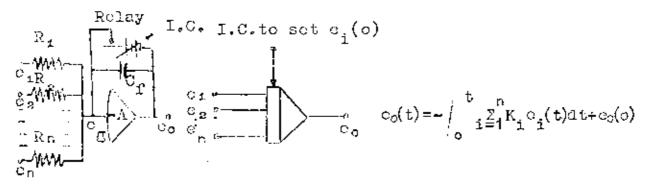
differential equations are shown in Fig.2.8, together with the corresponding mathematical functions.



d) Single Input Integrator



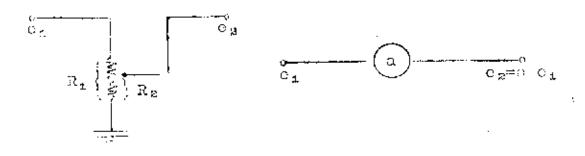
e) Single Input Integrator with I.C.

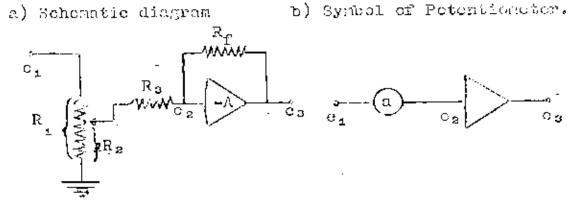


f) Multiple Inputs Integrator with L.C.

- <u>Pig.2.8</u>: Computer Diagrams, Symbols and Mathematical Functions of the Operational Amplifier.
  - (2.2) <u>Coefficient Setting Potentiometers</u>.<sup>(6)</sup>

Coefficient Setting Potentiometers are used in an analogue computer setups to perform multiplication by a constant less than unity. There are necessary for setting the coefficients of equations, and the potentioseter carcuit is shown in Fig.2.9.



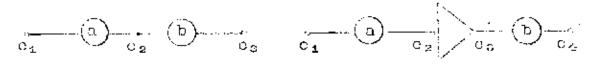


c) Loading of a Potentiometer and Symbol. <u>Pig.2.5</u>: The Coefficient Setting Potentiometer. In Fig.2.9a, we have  $c_2 = \frac{R_2}{R_1} c_1$ .

Since  $R_2 \leq R_1$ , the ratio  $\frac{R_2}{R_1}$ , denoted by a, is less than 1.

Hence  $c_2 = a c_1$ , where  $a \leq 1$ . ----- (2.9)

In practice, a potentiometer is usually connected as shown in Mig.2.9c. Two coefficient setting potentiometers of equal resistance cannot be used directly in series without an intervening operational amplifier (see Mig.2.10).



a) Not Possible. b) Possible.

Piz.2.10: The Method of Using Two Botentiemeters.

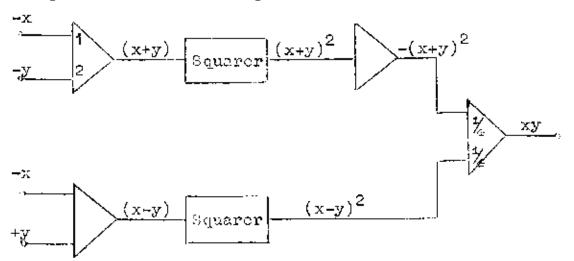
# (2.3) <u>Huldiplier and Munction Concrator</u>.<sup>(7)</sup>

### (c) <u>Auarter-Square Sultiplier</u>.

Nultiplication of two variable voltages is a nonlinear operation which is necessary on a general purpose computer. A "quarter-square" technique is used to effect this operation, use being made of the identity:

$$xy = \frac{1}{4} [(x+y)^2 - (x-y)^3] - (2.10)$$

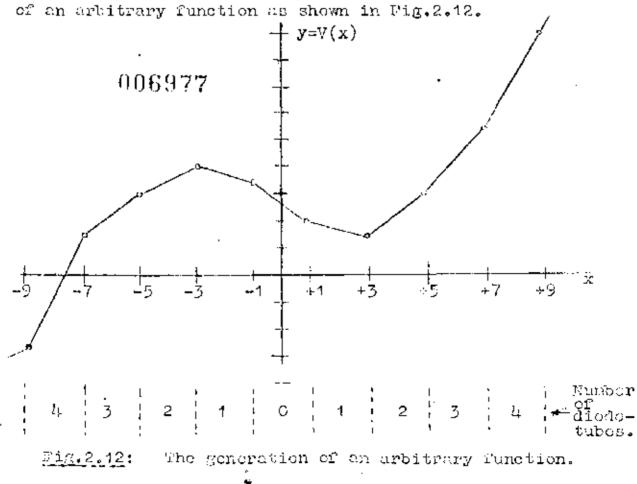
The block diagram representation of a quarter-square multiplior is shown in Fig.2.11.



212.2.11: Block Diagram of a Quarter-square Multiplier

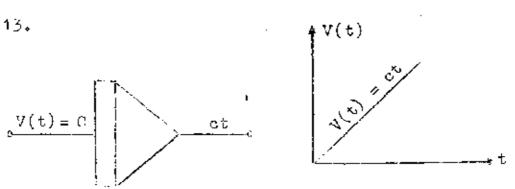
#### (b) Function Generator

The function generators available for an analogue computer use are of various types. In general, function generator in an analogue computer is done by the use of straight-line segments which are combined to approximate arbitrary curves. The functions are produced by the use of diode-tubes and each diode-tube makes a line segment. The number of straight-line segments usable in the representation normally varies from 5 to 22, depending upon the manufacturer of the equipment. An example for the generation of an arbitrary function as shown in Die 2 42



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Sometimes the computer does not have function generators, or the function generator of the computer connot construct the required function. Other techniques may be used to construct functions. For example a linear function of t may be constructed by supplying a constant potential to the input of an integrator. The output of the lintegrator is represented by a straight line, as shown in Fig.2.13.  $\star V(t)$ 



<u>Fig.2.13</u>: The method for constructing a linear function of t. Furthermore, we may use a signal generator<sup>(a)</sup> for the function generator of the computer. But the method of using the signal generator will not be discussed in this thesis.

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